



# Requirements and challenges for SDR implementation

Christer Svensson Linköping University

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#### Outline

The software defined radio Critical requirements Receiver solutions? Transmitter solutions? Conclusions





Software defined radio liberates radio-based services from chronic dependency on hard-wired characteristics including frequency band, channel bandwidth and channel coding.

(Joe Mitola, IEEE Communications Magazine, 2005)

Main motivation is that the number of wireless standards and utilized bands increase continuously. And, future Cognitive radio etc.





#### **Typical needs:**

Consumer terminal (radio, TV, cell phone, WiFi, GPS) carrier 100MHz-6GHz, bandwidth 200kHz-20MHz

Consumer terminal (cell phone, WiFi, GPS) carrier 700MHz-6GHz, bandwidth 200kHz-20MHz (LTE only 698-2690 MHz)

Military, Security (blue light authorities) carrier 2MHz-2GHz, 70MHz-400MHz





Key Issue

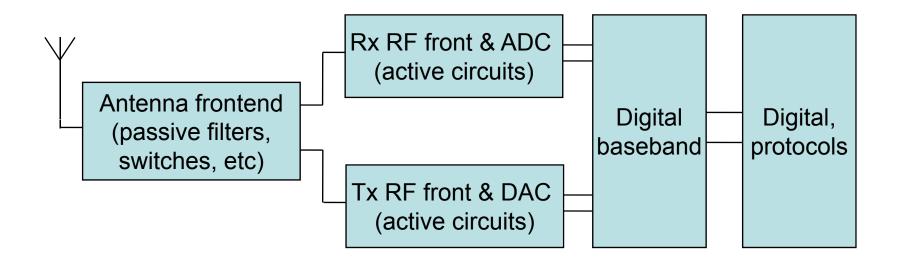
Can this be done?

If so, what technologies are available and which are lacking?





#### **Generic hardware architecture**







What have we?

Digital, protocols- a lot! Runs on GPP

Digital baseband - OK. Runs on DSP/FPGA or appl. Spec. DSPs

DSP/FPGA high power, application specific DSP low power (Nilsson)

Receiver analog frontend – multiband technology exists Insufficient

Receiver antenna frontend – multiband solutions lacking (Except very high cost military)

Transmitter PA and antenna frontend – multiband solutions lacking





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Receiver ante<br/>(Except very hPossibly a showstoppersckingTransmitter FWhy so little efforts here?<br/>olutions lacking





Unfortunately, we must adapt to reality.

**Receiver:** 

We need to receive a *weak signal* in presence of a *strong one* 

Best sensitivity – adapt to thermal noise background

Unintentional disturber (broadcast, nearby client, ...) Our own transmitter Intentional disturber (jammer)





A disturber transmitting power  $P_D$  at distance R with antenna gain  $G_D$  give a received (blocker) power in receiver with antenna gain  $G_r$ :

$$P_B = \frac{\lambda^2}{16\pi R^2} G_D G_r P_D$$

Where  $\lambda$  is the wavelength ( $\lambda = c/f_c$ ).

A blocker of power  $P_B$  gives a peak-to-peak voltage over  $R_0=50\Omega$  of

$$V_{p-p} = \sqrt{8P_BR_0}$$





**Best sensitivity – environmental noise at ambient temperature:** 

Noise spectral density:  $S_N = kT$ 

With a channel bandwidth of B, we require a dynamic range of:

$$DR = \frac{P_B}{S_N B}$$





This can be expressed as a ADC requirement of (quantization noise = thermal noise; f<sub>s</sub> sampling rate, n bits):

$$f_s 2^{2n} = \frac{4}{3} \frac{P_B}{kT}$$

Theory of ADC power consumption estimates ADC power to about  $30P_s$ , where  $P_s$  is the power needed to sample the signal (Sundström):

$$P_S = 24kTf_s 2^{2n} = 32P_B$$





#### We need to understand blocker power levels in practice

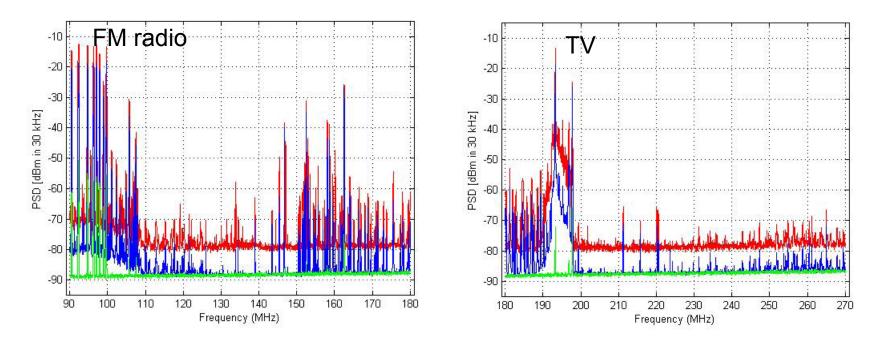
Direct measurements

Estimations for specific transmitters

Own transmitter







#### Measured background

Examples of Spectrum Occupancy measurements (Ellingson 2005) Max power about -12dBm. Above 1GHz frequencies -30dBm



#### **Specific transmitters**

Estimated power levels from specific transmitters

$$P_B = \frac{\lambda^2}{16\pi R^2} G_D G_r P_D$$

	f <sub>c</sub> , MHz	R, m	P <sub>T</sub> , W	G <sub>T</sub> , dBi	P <sub>B</sub> , dBm
VHF	70	3	10	2	30
Tetra	400	3	25	2	19
FM broadcast	90	300	50,000	-20	3
TV broadcast	400	300	50,000	-20	-10
GSM basest.	900	30	100	-20	-24
GSM terminal	900	2	1	2	1.5
WLAN	2400	2	1	2	-7



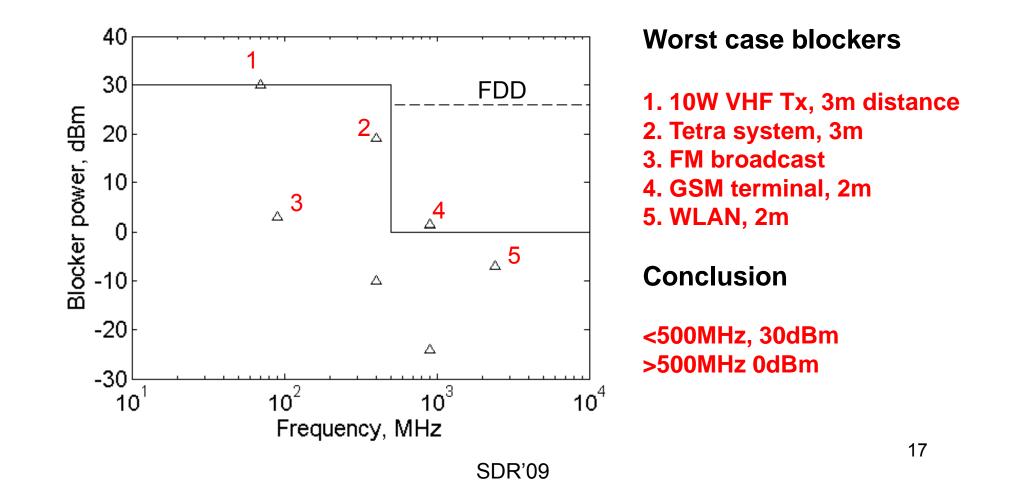


**Own transmitter (FDD)** 

Example 3G mobile, 26dBm, if same antenna  $P_B$ =26dBm











#### Consequences

Blocker	Voltage p-p	ADC power
1mW (0dBm)	0.63V	0.96W
1W (30dBm)	20V	960W

We can hardly accept more than ~1mW at ADC (ADC power) We can hardly accept more than ~1mW by receiver electronics (voltage)

#### This may be a show-stopper

Is any other contribution at this conference addressing this issue?





**Transmitter:** 

High power, 300mW - 30W, requires voltages of 11-110V @ $50\Omega$  May require special technology (except possibly 300mW)

Low spurious content

**High efficiency** Particularly tough at advanced modulation (including non-constant envelope)

#### Not available for wide frequency ranges





Very wide frequency range Passive tunable filters not availible No passive filters in frontend preferred

If not possible – we need tunable passive filters or filter banks

#### So, what is possible?

For blocker < 1mW maybe no passive filters possible

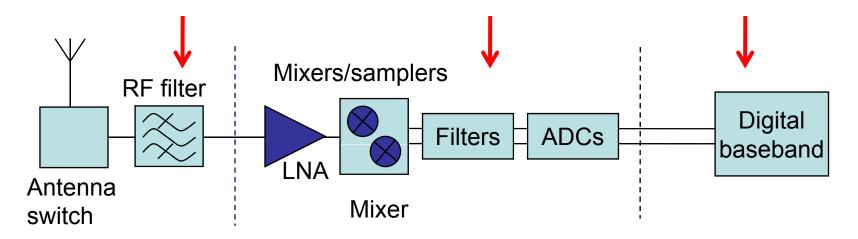
For blocker > 1mW passive filters mandatory





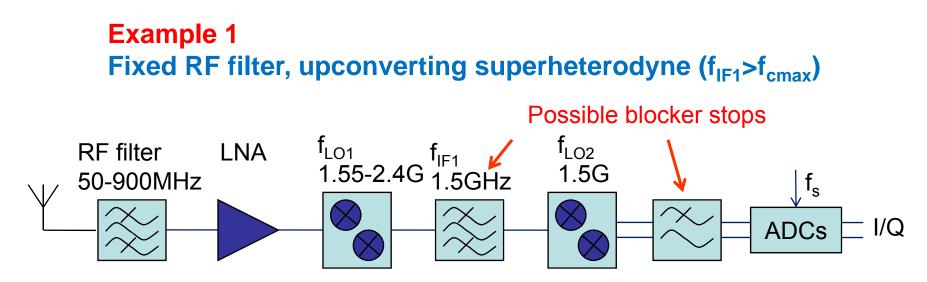
Blockers > 1mW MUST be stopped here NEED passive RF filters *unflexible <500MHz; FDD systems Tunable RF filter* 

Blockers ~1mW Stopped in digital baseband OR in analog baseband >500MHz, no FDD Fixed (all bands) RF filter









1dBm blocker at ADC: We can choose f<sub>s</sub>=100MHz gives n=16 (commercially available, eg. 160MS/s 16b 1.45W). Narrow IF and baseband may remove blockers; much relaxed ADC

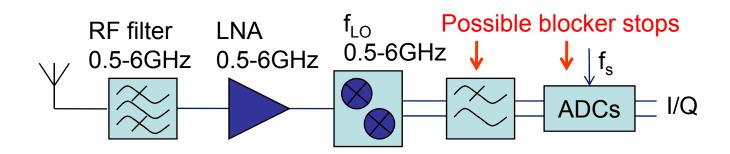
Similar solutions are available as TV receivers 50-900MHz, single chip complete receiver, ~1W (but P<sub>B</sub><-10dBm)





#### Example 2

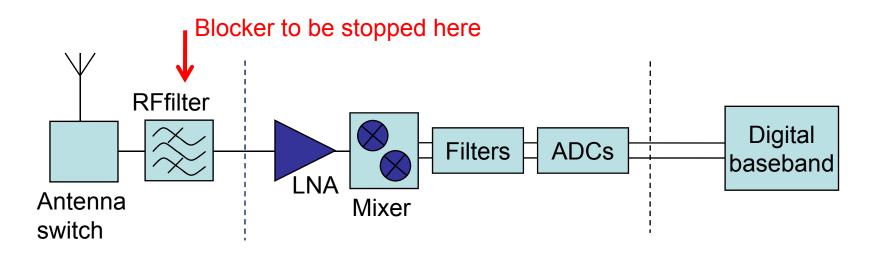
#### Fixed RF filter, high compression point LNA+mixer



Gain to be supplied by low-pass filter or ADC Note, LNA *and* mixer *and* ADC must have low noise figure! No solution at hand, but promising research (Rodriguez, Ahsan)







Need narrow band passive RF filters at input. Requirements: 1W blocker power Blocker offset frequency ~6% (military spec., CDMA FDD spec.)

Need >30dB attenuation at 6% frequency offset

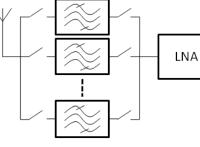


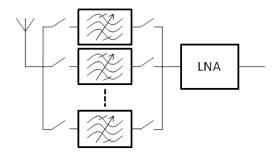


#### **Filter solutions**

#### **Filter bank**

Each filter 3% bandwidth and 3% allocated for slope. Requires 24 filters/octave. 100MHz-6GHz requires144 filters.





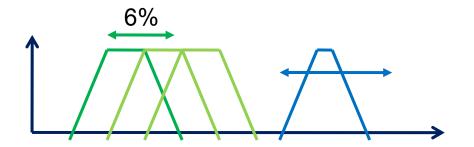
a) Fixed filter bank, 24 filters/octave

b) Tunable filter bank, 1 filter/octave

#### **Tunable filters**

1% bandwidth and 30dB attenuation at 6% requires second order.

1 octave tuning range 100MHz-6GHz requires 6 filters.





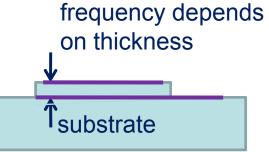
Filter candidates - Filter banks

#### Bulk acoustic wave filters (BAW, FBAR)

Presently used in mobile phones Good performance, small size (100µm) Unsuitable for frequencies below 500MHz Unclear how to fabricate 144 different frequencies High loss in switches

#### **MEMS** resonators

Good performance, small size (30-300 $\mu$ m) High impedances (>k $\Omega$ ) Insufficient linearity IIP<sub>3</sub><20dBm







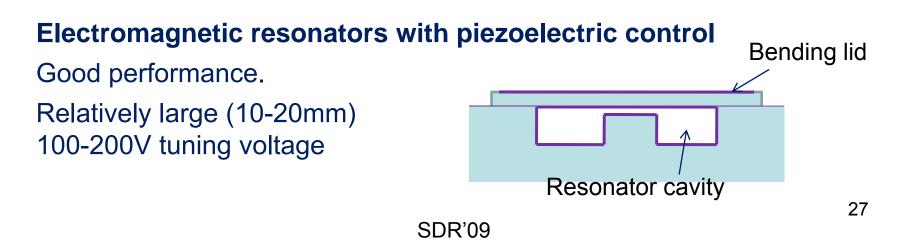


Filter candidates – Tunable filters

#### **Electromagnetic resonators with paraelectric varactor**

Commercially available (for antenna tuning) Sufficient performance.

Relatively large (10-20mm) 10-100V tuning voltage







**Conclusions narrow filters** 

To manage high blocking powers we need passive filters

#### Filter banks require very many filters

A tentatively possible solution is BAW filters

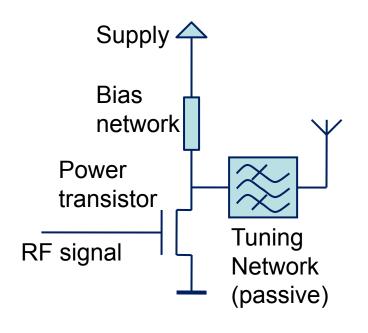
Tunable filters need much fewer filters, but they are larger

Tentative solutions are electromagnetic resonatorsbased onparaelectric varactorsorpiezoelectrically activated mechanical tuning





### **Transmitter solutions?**



Class A, AB, B May avoid tuning network Medium efficiency Low efficiency at lower power

Class C, E, F Must have tuning network Fixed output power (from fixed supply) Good efficiency

Class D May avoid tuning network Good efficiency possible



# Conclusions



#### Antenna/RF frontends are the most challenging part for SDR

Problems related to radio basics - not to waveforms or software

# **No technology available today for the most demanding requirements** >0dBm disturbers

Need tunable passive filters or switched passive filter banks

#### Some mobile/WiFi may be designed using available technology

≤ 0dBm disturbers, no FFD

Challenges: appropriate receiver architecture (learn from TV tuners) receiver dynamic range and linearity low power digital baseband (application specific DSP) appropriate PA technology





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